

## Super-Resolution



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# Outline

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- Classical methods for parameter estimation
  - Polynomial method: Prony's method
  - Subspace method: MUSIC
  - Matrix pencil algorithm
- Optimization-based methods
  - Basis mismatch
  - Atomic norm minimization
  - Connections to low-rank matrix completion

# Parameter estimation

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**Model:** a signal is mixture of  $r$  modes

$$x[t] = \sum_{i=1}^r d_i \psi(t; f_i), \quad t \in \mathbb{Z}$$

- $d_i$  : amplitudes
- $f_i$  : frequencies
- $\psi$ : (known) modal function, e.g.  $\psi(t, f_i) = e^{j2\pi t f_i}$
- $r$ : model order
- $2r$  unknown parameters:  $\{d_i\}$  and  $\{f_i\}$

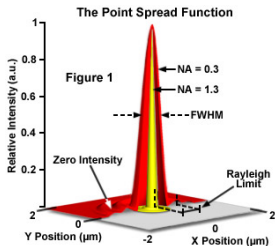
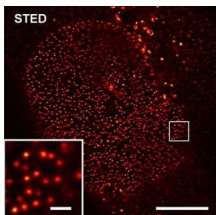
# Application: super-resolution imaging

Consider a time signal (a dual problem)

$$z(t) = \sum_{i=1}^r d_i \delta(t - t_i)$$

- Resolution is limited by point spread function  $h(t)$  of imaging system

$$x(t) = z(t) * h(t)$$



point spread function  $h(t)$

## Application: super-resolution imaging

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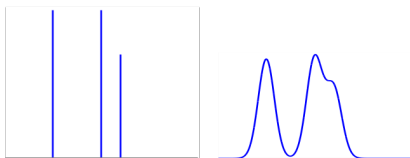
$$\text{time domain: } x(t) = z(t) * h(t) = \sum_{i=1}^r d_i h(t - t_i)$$

$$\text{spectral domain: } \hat{x}(f) = \hat{z}(f) \hat{h}(f) = \sum_{i=1}^r d_i \underbrace{\hat{h}(f)}_{\text{known}} e^{j2\pi f t_i}$$

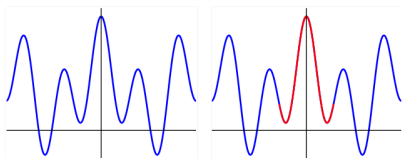
$$\implies \text{observed data: } \frac{\hat{x}(f)}{\hat{h}(f)} = \sum_{i=1}^r d_i \underbrace{e^{j2\pi f t_i}}_{\psi(f; t_i)}, \quad \forall f : \hat{h}(f) \neq 0$$

$h(t)$  is usually band-limited (suppress high-frequency components)

# Application: super-resolution imaging



(a) highly resolved signal  $z(t)$ ; (b) low-pass version  $x(t)$



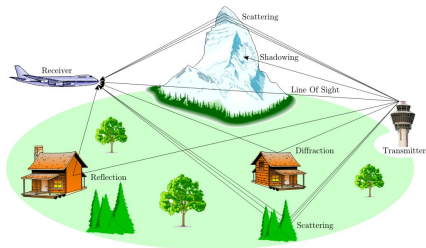
(c) Fourier transform  $\hat{z}(f)$ ; (d) (red) observed spectrum  $\hat{x}(f)$

Fig. credit: Candes, Fernandez-Granda '14

**Super-resolution:** extrapolate high-end spectrum (fine scale details)  
from low-end spectrum (low-resolution data)

# Application: multipath communication channels

In wireless communications, transmitted signals typically reach receiver by multiple paths, due to reflection from objects (e.g. buildings).



Suppose  $h(t)$  is transmitted signal, then received signal is

$$x(t) = \sum_{i=1}^r d_i h(t - t_i) \quad (t_i : \text{delay in } i^{\text{th}} \text{ path})$$

→ same as super-resolution model

# Basic model

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- **Signal model:** a mixture of sinusoids at  $r$  distinct frequencies

$$x[t] = \sum_{i=1}^r d_i e^{j2\pi t f_i}$$

where  $f_i \in [0, 1)$  : frequencies;  $d_i$  : amplitudes

- *Continuous dictionary:*  $f_i$  can assume ANY value in  $[0, 1)$

- **Observed data:**

$$\mathbf{x} = [x[0], \dots, x[n-1]]^\top$$

- **Goal:** retrieve frequencies / recover signal (also called **harmonic retrieval**)



# Matrix / vector representation

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Alternatively, observed data can be written as

$$\mathbf{x} = \mathbf{V}_{n \times r} \mathbf{d} \quad (15.1)$$

where  $\mathbf{d} = [d_1, \dots, d_r]^\top$ ;

$$\mathbf{V}_{n \times r} := \begin{bmatrix} 1 & 1 & 1 & \cdots & 1 \\ z_1 & z_2 & z_3 & \cdots & z_r \\ z_1^2 & z_2^2 & z_3^2 & \cdots & z_r^2 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ z_1^{n-1} & z_2^{n-1} & z_3^{n-1} & \cdots & z_r^{n-1} \end{bmatrix} \quad (\text{Vandermonde matrix})$$

with  $z_i = e^{j2\pi f_i}$ .

# Polynomial method: Prony's method

# Prony's method

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- A *polynomial method* proposed by Gaspard Riche de Prony in 1795
- **Key idea:** construct an annihilating filter + polynomial root finding

# Annihilating filter

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- Define a filter by (**Z-transform** or **characteristic polynomial**)

$$G(z) = \sum_{l=0}^r g_l z^{-l} = \prod_{l=1}^r (1 - z_l z^{-1})$$

whose roots are  $\{z_l = e^{j2\pi f_l} \mid 1 \leq l \leq r\}$

- $G(z)$  is called **annihilating filter** since it annihilates  $x[k]$ , i.e.

$$q[k] := \underbrace{g_k * x[k]}_{\text{convolution}} = 0 \quad (15.2)$$

**Proof:**

$$\begin{aligned} q[k] &= \sum_{i=0}^r g_i x[k-i] = \sum_{i=0}^r \sum_{l=1}^r g_i d_l z_l^{k-i} \\ &= \sum_{l=1}^r d_l z_l^k \left( \underbrace{\sum_{i=0}^r g_i z_l^{-i}}_{=0} \right) = 0 \end{aligned}$$

# Annihilating filter

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Equivalently, one can write (15.2) as

$$\mathbf{X}_e \mathbf{g} = \mathbf{0}, \quad (15.3)$$

where  $\mathbf{g} = [g_r, \dots, g_0]^\top$  and

$$\mathbf{X}_e := \underbrace{\begin{pmatrix} x[0] & x[1] & x[2] & \cdots & x[r] \\ x[1] & x[2] & x[3] & \cdots & x[r+1] \\ x[2] & x[3] & x[4] & \cdots & x[r+2] \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ x[n-r-1] & x[n-r] & \cdots & \cdots & x[n-1] \end{pmatrix}}_{\text{Hankel matrix}} \in \mathbb{C}^{(n-r) \times (r+1)} \quad (15.4)$$

Thus, we can obtain coefficients  $\{g_i\}$  by solving linear system (15.3).

# A crucial decomposition

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## Vandermonde decomposition

$$\mathbf{X}_e = \mathbf{V}_{(n-r) \times r} \text{diag}(\mathbf{d}) \mathbf{V}_{(r+1) \times r}^\top \quad (15.5)$$

**Implications:** if  $n \geq 2r$  and  $d_i \neq 0$ , then

- $\text{rank}(\mathbf{X}_e) = \text{rank}(\mathbf{V}_{(n-r) \times r}) = \text{rank}(\mathbf{V}_{(r+1) \times r}) = r$
- $\text{null}(\mathbf{X}_e)$  is 1-dimensional  $\iff$  nonzero solution to  $\mathbf{X}_e \mathbf{g} = \mathbf{0}$  is unique

# A crucial decomposition

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## Vandermonde decomposition

$$\mathbf{X}_e = \mathbf{V}_{(n-r) \times r} \text{diag}(\mathbf{d}) \mathbf{V}_{(r+1) \times r}^\top \quad (15.5)$$

**Proof:** For any  $i$  and  $j$ ,

$$\begin{aligned} [\mathbf{X}_e]_{i,j} &= x[i+j-2] = \sum_{l=1}^r d_l z_l^{i+j-2} = \sum_{l=1}^r z_l^{i-1} d_l z_l^{j-1} \\ &= \left( \mathbf{V}_{(n-r) \times r} \right)_{i,:} \text{diag}(\mathbf{d}) \left( \mathbf{V}_{(r+1) \times r} \right)_{j,:}^\top \end{aligned}$$

# Prony's method

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## Algorithm 15.1 Prony's method

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- 1 Find  $\mathbf{g} = [g_r, \dots, g_0]^T \neq \mathbf{0}$  that solves  $\mathbf{X}_e \mathbf{g} = \mathbf{0}$
  - 2 Compute  $r$  roots  $\{z_l \mid 1 \leq l \leq r\}$  of  $G(z) = \sum_{l=0}^r g_l z^{-l}$
  - 3 Calculate  $f_l$  via  $z_l = e^{j2\pi f_l}$
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## Drawback

- Root-finding for polynomials becomes difficult for large  $r$
- Numerically unstable in the presence of noise



## Subspace method: MUSIC

# MULTIPLE Signal Classification (MUSIC)

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Consider a (slightly more general) Hankel matrix

$$\mathbf{X}_e = \begin{pmatrix} x[0] & x[1] & x[2] & \cdots & x[k] \\ x[1] & x[2] & x[3] & \cdots & x[k+1] \\ x[2] & x[3] & x[4] & \cdots & x[k+2] \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ x[n-k-1] & x[n-k] & \cdots & \cdots & x[n-1] \end{pmatrix} \in \mathbb{C}^{(n-k) \times (k+1)}$$

where  $r \leq k \leq n - r$  (note that  $k = r$  in Prony's method).

- $\text{null}(\mathbf{X}_e)$  might span multiple dimensions

# MULTiple Signal Classification (MUSIC)

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- Generalize Prony's method by computing  $\{\mathbf{v}_i \mid 1 \leq i \leq k - r + 1\}$  that forms orthonormal basis for  $\text{null}(\mathbf{X}_e)$

- Let  $\mathbf{z}(f) := \begin{bmatrix} 1 \\ e^{j2\pi f} \\ \vdots \\ e^{j2\pi k f} \end{bmatrix}$ , then it follows from Vandermonde decomposition that

$$\mathbf{z}(f_l)^\top \mathbf{v}_i = 0, \quad 1 \leq i \leq k - r + 1, \quad 1 \leq l \leq r$$

- Thus,  $\{f_l\}$  are **peaks** in pseudospectrum

$$S(f) := \frac{1}{\sum_{i=1}^{k-r+1} |\mathbf{z}(f)^\top \mathbf{v}_i|^2}$$

# MUSIC algorithm

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## Algorithm 15.2 MUSIC

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- 1 Compute orthonormal basis  $\{\mathbf{v}_i \mid 1 \leq i \leq k - r + 1\}$  for  $\text{null}(\mathbf{X}_e)$
  - 2 Return  $r$  largest peaks of  $S(f) := \frac{1}{\sum_{i=1}^{k-r+1} |\mathbf{z}(f)^\top \mathbf{v}_i|^2}$ , where  $\mathbf{z}(f) := [1, e^{j2\pi f}, \dots, e^{j2\pi k f}]^\top$
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# Matrix pencil algorithm

# Matrix pencil

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## Definition 15.1

A *linear matrix pencil*  $M(\lambda)$  is defined as a combination

$$M(\lambda) = M_1 - \lambda M_2$$

of 2 matrices  $M_1$  and  $M_2$ , where  $\lambda \in \mathbb{C}$ .

# Matrix pencil algorithm

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## Key idea:

- Create 2 closely related Hankel matrices  $\mathbf{X}_{e,1}$  and  $\mathbf{X}_{e,2}$
- Recover  $\{f_l\}$  via generalized eigenvalues of matrix pencil

$$\mathbf{X}_{e,2} - \lambda \mathbf{X}_{e,1}$$

(i.e. all  $\lambda \in \mathbb{C}$  obeying  $\det(\mathbf{X}_{e,2} - \lambda \mathbf{X}_{e,1}) = 0$ )

# Matrix pencil algorithm

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Construct 2 Hankel matrices (which differ only by 1 row)

$$\mathbf{X}_{e,1} := \begin{pmatrix} x[0] & x[1] & x[2] & \cdots & x[k-1] \\ x[1] & x[2] & x[3] & \cdots & x[k] \\ x[2] & x[3] & x[4] & \cdots & x[k+1] \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ x[n-k-1] & x[n-k] & \cdots & \cdots & x[n-2] \end{pmatrix} \in \mathbb{C}^{(n-k) \times k}$$
$$\mathbf{X}_{e,2} := \begin{pmatrix} x[1] & x[2] & x[3] & \cdots & x[k] \\ x[2] & x[3] & x[4] & \cdots & x[k+1] \\ x[3] & x[4] & x[5] & \cdots & x[k+2] \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ x[n-k] & x[n-k+1] & \cdots & \cdots & x[n-1] \end{pmatrix} \in \mathbb{C}^{(n-k) \times k}$$

where  $k$  is pencil parameter



# Matrix pencil algorithm

## Fact 15.2

Similar to (15.5), one has

$$\mathbf{X}_{e,1} = \mathbf{V}_{(n-k) \times r} \text{diag}(\mathbf{d}) \mathbf{V}_{k \times r}^\top$$

$$\mathbf{X}_{e,2} = \mathbf{V}_{(n-k) \times r} \text{diag}(\mathbf{d}) \text{diag}(\mathbf{z}) \mathbf{V}_{k \times r}^\top$$

where  $\mathbf{z} = [e^{j2\pi f_1}, \dots, e^{j2\pi f_r}]^\top$ .

**Proof:** The result for  $\mathbf{X}_{e,1}$  follows from (15.5). Regarding  $\mathbf{X}_{e,2}$ ,

$$\begin{aligned} [\mathbf{X}_{e,2}]_{i,j} &= x[i+j-1] = \sum_{l=1}^r d_l z_l^{i+j-1} = \sum_{l=1}^r z_l^{i-1} (d_l z_l) z_l^{j-1} \\ &= (\mathbf{V}_{(n-k) \times r})_{i,:} \text{diag}(\mathbf{d}) \text{diag}(\mathbf{z}) (\mathbf{V}_{k \times r})_{j,:}^\top \end{aligned}$$

# Matrix pencil algorithm

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Based on Fact 15.2,

$$\mathbf{X}_{e,2} - \lambda \mathbf{X}_{e,1} = \mathbf{V}_{(n-k) \times r} \text{diag}(\mathbf{d}) (\text{diag}(\mathbf{z}) - \lambda \mathbf{I}) \mathbf{V}_{k \times r}^\top$$

(Exercise) If  $r \leq k \leq n - r$ , then

- $\text{rank}(\mathbf{V}_{(n-k) \times r}) = \text{rank}(\mathbf{V}_{k \times r}) = r$
- Generalized eigenvalues of  $\mathbf{X}_{e,2} - \lambda \mathbf{X}_{e,1}$  are  $\{z_l = e^{j2\pi f_l}\}$ , which can be obtained by finding eigenvalues of  $\mathbf{X}_{e,1}^\dagger \mathbf{X}_{e,2}$

# Matrix pencil algorithm

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## Algorithm 15.3 Matrix pencil algorithm

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- 1 Compute all eigenvalues  $\{\lambda_l\}$  of  $\mathbf{X}_{e,1}^\dagger \mathbf{X}_{e,2}$
  - 2 Calculate  $f_l$  via  $\lambda_l = e^{j2\pi f_l}$
-

## **Basis mismatch issue**

# Optimization methods for super resolution?

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Recall our representation in (15.1):

$$\mathbf{x} = \mathbf{V}_{n \times r} \mathbf{d} \quad (15.6)$$

- **Challenge:** both  $\mathbf{V}_{n \times r}$  and  $\mathbf{d}$  are **unknown**

Alternatively, one can view (15.6) as sparse representation over a **continuous** dictionary  $\{\mathbf{z}(f) \mid 0 \leq f < 1\}$ , where  $\mathbf{z}(f) = [1, e^{j2\pi f}, \dots, e^{j2\pi(n-1)f}]^\top$

# Optimization methods for super resolution?

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## One strategy:

- Convert nonlinear representation into linear system via discretization at desired resolution:

$$\text{(assume)} \quad \mathbf{x} = \underbrace{\Psi}_{n \times p \text{ partial DFT matrix}} \boldsymbol{\beta}$$

- representation over a discrete frequency set  $\{0, \frac{1}{p}, \dots, \frac{p-1}{p}\}$
- gridding resolution:  $1/p$
- Solve  $\ell_1$  minimization:

$$\text{minimize}_{\boldsymbol{\beta} \in \mathbb{C}^p} \|\boldsymbol{\beta}\|_1 \quad \text{s.t. } \mathbf{x} = \Psi \boldsymbol{\beta}$$

## Discretization destroys sparsity

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Suppose  $n = p$ , and recall

$$\begin{aligned} \mathbf{x} &= \Psi\boldsymbol{\beta} = \mathbf{V}_{n \times r}\mathbf{d} \\ \implies \boldsymbol{\beta} &= \Psi^{-1}\mathbf{V}_{n \times r}\mathbf{d} \end{aligned}$$

Ideally, if  $\Psi^{-1}\mathbf{V}_{n \times r} \approx$  submatrix of  $\mathbf{I}$ , then sparsity is preserved.

# Discretization destroys sparsity

---

Suppose  $n = p$ , and recall

$$\begin{aligned} \mathbf{x} &= \Psi\boldsymbol{\beta} = \mathbf{V}_{n \times r} \mathbf{d} \\ \implies \boldsymbol{\beta} &= \Psi^{-1} \mathbf{V}_{n \times r} \mathbf{d} \end{aligned}$$

Simple calculation gives

$$\Psi^{-1} \mathbf{V}_{n \times r} = \begin{bmatrix} D(\delta_0) & D(\delta_1) & \cdots & D(\delta_r) \\ D(\delta_0 - \frac{1}{p}) & D(\delta_1 - \frac{1}{p}) & \cdots & D(\delta_r - \frac{1}{p}) \\ \vdots & \vdots & \ddots & \vdots \\ D(\delta_0 - \frac{p-1}{p}) & D(\delta_1 - \frac{p-1}{p}) & \cdots & D(\delta_r - \frac{p-1}{p}) \end{bmatrix}$$

where  $f_i$  is mismatched to grid  $\{0, \frac{1}{p}, \dots, \frac{p-1}{p}\}$  by  $\delta_i$ , and

$$D(f) := \frac{1}{p} \sum_{l=0}^{p-1} e^{j2\pi lf} = \frac{1}{p} e^{j\pi f(p-1)} \underbrace{\frac{\sin(\pi fp)}{\sin(\pi f)}}_{\text{heavy tail}} \quad (\text{Dirichlet kernel})$$



# Discretization destroys sparsity

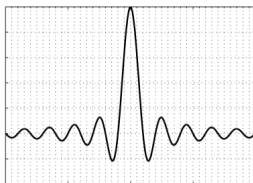
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Suppose  $n = p$ , and recall

$$\mathbf{x} = \Psi\boldsymbol{\beta} = \mathbf{V}_{n \times r}\mathbf{d}$$

$$\implies \boldsymbol{\beta} = \Psi^{-1}\mathbf{V}_{n \times r}\mathbf{d}$$

Slow decay / spectral leakage of Dirichlet kernel



If  $\delta_i = 0$  (no mismatch),  $\Psi^{-1}\mathbf{V}_{n \times r}$  = submatrix of  $\mathbf{I}$

$\implies \Psi^{-1}\mathbf{V}_{n \times r}\mathbf{d}$  is sparse

# Discretization destroys sparsity

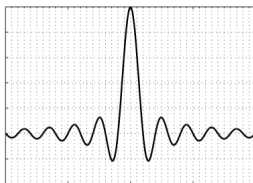
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Suppose  $n = p$ , and recall

$$\mathbf{x} = \Psi\boldsymbol{\beta} = \mathbf{V}_{n \times r}\mathbf{d}$$

$$\implies \boldsymbol{\beta} = \Psi^{-1}\mathbf{V}_{n \times r}\mathbf{d}$$

Slow decay / spectral leakage of Dirichlet kernel



If  $\delta_i \neq 0$  (e.g. randomly generated),  $\Psi^{-1}\mathbf{V}_{n \times r}$  may be far from submatrix of  $\mathbf{I}$

$\implies \Psi^{-1}\mathbf{V}_{n \times r}\mathbf{d}$  may be incompressible

- Finer gridding does not help!

# Mismatch of DFT basis

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Loss of sparsity after discretization due to basis mismatch

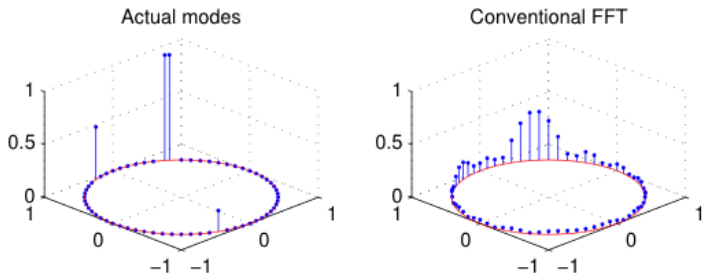


Fig. credit: Chi, Pezeshki, Scharf, Calderbank '10

# Optimization-based methods

# Atomic set

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Consider a set of atoms  $\mathcal{A} = \{\psi(\nu) : \nu \in S\}$

- $S$  can be finite, countable, or even continuous
- **Examples:**
  - standard basis vectors (used in compressed sensing)
  - rank-one matrices (used in low-rank matrix recovery)
  - line spectral atoms

$$a(f, \phi) := \underbrace{e^{j\phi}}_{\text{global phase}} [1, e^{j2\pi f}, \dots, e^{j2\pi(n-1)f}]^\top, \quad f \in [0, 1), \quad \phi \in [0, 2\pi)$$

# Atomic norm

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## Definition 15.3 (Atomic norm, Chandrasekaran, Recht, Parrilo, Willsky '10)

Atomic norm of any  $\mathbf{x}$  is defined as

$$\|\mathbf{x}\|_{\mathcal{A}} := \inf \left\{ \|\mathbf{d}\|_1 : \mathbf{x} = \sum_k d_k \psi(\nu_k) \right\} = \inf \{t > 0 : \mathbf{x} \in t \operatorname{conv}(\mathcal{A})\}$$

- Generalization of  $\ell_1$  norm for vectors

# SDP representation of atomic norm

Consider set of line spectral atoms

$\mathcal{A} := \left\{ a(f, \phi) := e^{j\phi} \cdot [1, e^{j2\pi f}, \dots, e^{j2\pi(n-1)f}]^\top \mid f \in [0, 1), \phi \in [0, 2\pi) \right\}$ ,  
then

$$\|\mathbf{x}\|_{\mathcal{A}} = \inf_{d_k \geq 0, \phi_k \in [0, 2\pi), f_k \in [0, 1)} \left\{ \sum_k d_k \mid \mathbf{x} = \sum_k d_k a(f_k, \phi_k) \right\}$$

## Lemma 15.4 (Tang, Bhaskar, Shah, Recht '13)

For any  $\mathbf{x} \in \mathbb{C}^n$ ,

$$\|\mathbf{x}\|_{\mathcal{A}} = \inf \left\{ \frac{1}{2n} \text{Tr}(\text{Toeplitz}(\mathbf{u})) + \frac{1}{2}t \mid \begin{bmatrix} \text{Toeplitz}(\mathbf{u}) & \mathbf{x} \\ \mathbf{x}^* & t \end{bmatrix} \succeq \mathbf{0} \right\} \quad (15.7)$$

# Vandermonde decomposition of PSD Toeplitz matrices

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## Lemma 15.5

Any Toeplitz matrix  $\mathbf{P} \succeq \mathbf{0}$  can be represented as

$$\mathbf{P} = \mathbf{V} \text{diag}(\mathbf{d}) \mathbf{V}^*,$$

where  $\mathbf{V} := [a(f_1, 0), \dots, a(f_r, 0)]$ ,  $d_i \geq 0$ , and  $r = \text{rank}(\mathbf{P})$

- Vandermonde decomposition can be computed efficiently via root finding



## Proof of Lemma 15.4

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Let  $\text{SDP}(\mathbf{x})$  be value of RHS of (15.7).

1. **Show that**  $\text{SDP}(\mathbf{x}) \leq \|\mathbf{x}\|_{\mathcal{A}}$ .

- Suppose  $\mathbf{x} = \sum_k d_k a(f_k, \phi_k)$  for  $d_k \geq 0$ . Picking  $\mathbf{u} = \sum_k d_k a(f_k, 0)$  and  $t = \sum_k d_k$  gives (exercise)

$$\text{Toeplitz}(\mathbf{u}) = \sum_k d_k a(f_k, 0) a^*(f_k, 0) = \sum_k d_k a(f_k, \phi_k) a^*(f_k, \phi_k)$$

$$\Rightarrow \begin{bmatrix} \text{Toeplitz}(\mathbf{u}) & \mathbf{x} \\ \mathbf{x}^* & t \end{bmatrix} = \sum_k d_k \begin{bmatrix} a(f_k, \phi_k) \\ 1 \end{bmatrix} \begin{bmatrix} a(f_k, \phi_k) \\ 1 \end{bmatrix}^* \succeq \mathbf{0}$$

- Given that  $\frac{1}{n} \text{Tr}(\text{Toeplitz}(\mathbf{u})) = t = \sum_k d_k$ , one has

$$\text{SDP}(\mathbf{x}) \leq \sum_k d_k.$$

Since this holds for any decomposition of  $\mathbf{x}$ , we conclude this part.

## Proof of Lemma 15.4

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2. **Show that**  $\|\mathbf{x}\|_{\mathcal{A}} \leq \text{SDP}(\mathbf{x})$ .

i) Suppose for some  $\mathbf{u}$ ,

$$\begin{bmatrix} \text{Toeplitz}(\mathbf{u}) & \mathbf{x} \\ \mathbf{x}^* & t \end{bmatrix} \succeq \mathbf{0}. \quad (15.8)$$

Lemma 15.5 suggests Vandermonde decomposition

$$\text{Toeplitz}(\mathbf{u}) = \mathbf{V} \text{diag}(\mathbf{d}) \mathbf{V}^* = \sum_k d_k a(f_k, 0) a^*(f_k, 0).$$

This together with the fact  $\|a(f_k, 0)\| = \sqrt{n}$  gives

$$\frac{1}{n} \text{Tr}(\text{Toeplitz}(\mathbf{u})) = \sum_k d_k.$$

## Proof of Lemma 15.4

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2. Show that  $\|\mathbf{x}\|_{\mathcal{A}} \leq \text{SDP}(\mathbf{x})$ .

ii) It follows from (15.8) that  $\mathbf{x} \in \text{range}(\mathbf{V})$ , i.e.

$$\mathbf{x} = \sum_k w_k a(f_k, 0) = \mathbf{V}\mathbf{w}$$

for some  $\mathbf{w}$ . By Schur complement lemma,

$$\mathbf{V}\text{diag}(\mathbf{d})\mathbf{V}^* \succeq \frac{1}{t}\mathbf{x}\mathbf{x}^* = \frac{1}{t}\mathbf{V}\mathbf{w}\mathbf{w}^*\mathbf{V}^*.$$

Let  $\mathbf{q}$  be any vector s.t.  $\mathbf{V}^*\mathbf{q} = \text{sign}(\mathbf{w})$ . Then

$$\begin{aligned}\sum_k d_k &= \mathbf{q}^*\mathbf{V}\text{diag}(\mathbf{d})\mathbf{V}^*\mathbf{q} \succeq \frac{1}{t}\mathbf{q}^*\mathbf{V}\mathbf{w}\mathbf{w}^*\mathbf{V}^*\mathbf{q} = \frac{1}{t}\left(\sum_k |w_k|\right)^2 \\ &\Rightarrow t \sum_k d_k \geq \left(\sum_k |w_k|\right)^2\end{aligned}$$

$$\stackrel{\text{AM-GM inequality}}{\implies} \frac{1}{2n}\text{Tr}(\text{Toeplitz}(\mathbf{u})) + \frac{1}{2}t \geq \sqrt{t \sum_k d_k} \geq \sum_k |w_k| \geq \|\mathbf{x}\|_{\mathcal{A}}$$

# Atomic norm minimization

---

$$\begin{aligned} & \text{minimize}_{\mathbf{z} \in \mathbb{C}^n} \quad \|\mathbf{z}\|_{\mathcal{A}} \\ & \text{s.t.} \quad z_i = x_i, \quad i \in T \quad (\text{observation set}) \end{aligned}$$

$\Leftrightarrow$

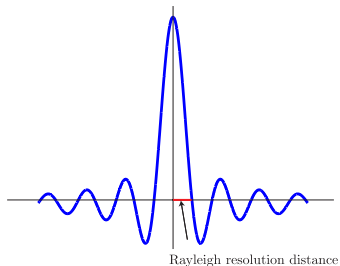
$$\begin{aligned} & \text{minimize}_{\mathbf{z} \in \mathbb{C}^n} \quad \frac{1}{2n} \text{Tr}(\text{Toeplitz}(\mathbf{u})) + \frac{1}{2}t \\ & \text{s.t.} \quad z_i = x_i, \quad i \in T \\ & \quad \quad \begin{bmatrix} \text{Toeplitz}(\mathbf{u}) & \mathbf{z} \\ \mathbf{z}^* & t \end{bmatrix} \succeq \mathbf{0} \end{aligned}$$

# Key metrics

---

**Minimum separation  $\Delta$**  of  $\{f_l \mid 1 \leq l \leq r\}$  is

$$\Delta := \min_{i \neq l} |f_i - f_l|$$



**Rayleigh resolution limit:**  $\lambda_c = \frac{2}{n-1}$

# Performance guarantees for super resolution

---

Suppose  $T = \{-\frac{n-1}{2}, \dots, \frac{n-1}{2}\}$

## Theorem 15.6 (Candes, Fernandez-Granda '14)

*Suppose that*

- **Separation condition:**  $\Delta \geq \frac{4}{n-1} = 2\lambda_c;$

*Then atomic norm (or total-variation) minimization is exact.*

- A deterministic result
- Can recover at most  $n/4$  spikes from  $n$  consecutive samples
- Does not depend on amplitudes / phases of spikes
- There is no separation requirement if all  $d_i$  are positive

# Fundamental resolution limits

---

If  $T = \{-\frac{n-1}{2}, \dots, \frac{n-1}{2}\}$ , we cannot go below Rayleigh limit  $\lambda_c$ .

## Theorem 15.7 (Moitra '15)

*If  $\Delta < \frac{2}{n-1} = \lambda_c$ , then no estimator can distinguish between a particular pair of  $\Delta$ -separated signals even under exponentially small noise.*

# Compressed sensing off the grid

---

Suppose  $T$  is **random** subset of  $\{0, \dots, N-1\}$  of cardinality  $n$   
— Extend compressed sensing to continuous domain

## Theorem 15.8 (Tang, Bhaskar, Shah, Recht '13)

*Suppose that*

- **Random sign:**  $\text{sign}(d_i)$  are *i.i.d. and random*;
- **Separation condition:**  $\Delta \geq \frac{4}{N-1}$ ;
- **Sample size:**  $n \gtrsim \max\{r \log r \log N, \log^2 N\}$ .

*Then atomic norm minimization is exact with high prob.*

- Random sampling improves resolution limits ( $\frac{4}{n-1}$  vs.  $\frac{4}{N-1}$ )



# Connection to low-rank matrix completion

---

Recall Hankel matrix

$$\mathbf{X}_e := \begin{pmatrix} x[0] & x[1] & x[2] & \cdots & x[k] \\ x[1] & x[2] & x[3] & \cdots & x[k+1] \\ x[2] & x[3] & x[4] & \cdots & x[k+2] \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ x[n-k-1] & x[n-k] & \cdots & \cdots & x[n-1] \end{pmatrix}$$
$$= \mathbf{V}_{(n-k) \times r} \text{diag}(\mathbf{d}) \mathbf{V}_{(k+1) \times r}^\top \quad (\text{Vandermonde decomposition})$$

- $\text{rank}(\mathbf{X}_e) \leq r$
- Spectral sparsity  $\iff$  low rank

# Recovery via Hankel matrix completion

---

Enhanced Matrix Completion (EMaC):

$$\begin{aligned} & \underset{z \in \mathbb{C}^n}{\text{minimize}} && \|Z_e\|_* \\ & \text{s.t.} && z_i = x_i, \quad i \in T \end{aligned}$$

When  $T$  is random subset of  $\{0, \dots, N-1\}$ :

- Coherence measure is closely related to separation condition (Liao & Fannjiang '16)
- Similar performance guarantees as atomic norm minimization (Chen, Chi, Goldsmith '14)

## Extension to 2D frequencies

---

**Signal model:** a mixture of 2D sinusoids at  $r$  distinct frequencies

$$x[\mathbf{t}] = \sum_{i=1}^r d_i e^{j2\pi \langle \mathbf{t}, \mathbf{f}_i \rangle}$$

where  $\mathbf{f}_i \in [0, 1)^2$  : frequencies;  $d_i$  : amplitudes

- Multi-dimensional model:  $\mathbf{f}_i$  can assume ANY value in  $[0, 1)^2$

# Vandermonde decomposition

---

$$\mathbf{X} = [x(t_1, t_2)]_{0 \leq t_1 < n_1, 0 \leq t_2 < n_2}$$

**Vandermonde decomposition:**

$$\mathbf{X} = \mathbf{Y} \cdot \text{diag}(\mathbf{d}) \cdot \mathbf{Z}^\top.$$

where

$$\mathbf{Y} := \begin{bmatrix} 1 & 1 & \cdots & 1 \\ y_1 & y_2 & \cdots & y_r \\ \vdots & \vdots & \vdots & \vdots \\ y_1^{n_1-1} & y_2^{n_1-1} & \cdots & y_r^{n_1-1} \end{bmatrix}, \mathbf{Z} := \begin{bmatrix} 1 & 1 & \cdots & 1 \\ z_1 & z_2 & \cdots & z_r \\ \vdots & \vdots & \vdots & \vdots \\ z_1^{n_2-1} & z_2^{n_2-1} & \cdots & z_r^{n_2-1} \end{bmatrix}$$

with  $y_i = \exp(j2\pi f_{1i})$ ,  $z_i = \exp(j2\pi f_{2i})$ .

# Multi-fold Hankel matrix (Hua '92)

---

An enhanced form  $\mathbf{X}_e$ :  $k_1 \times (n_1 - k_1 + 1)$  block Hankel matrix

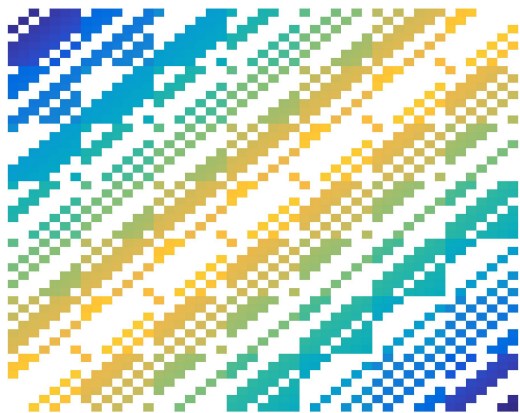
$$\mathbf{X}_e = \begin{bmatrix} \mathbf{X}_0 & \mathbf{X}_1 & \cdots & \mathbf{X}_{n_1-k_1} \\ \mathbf{X}_1 & \mathbf{X}_2 & \cdots & \mathbf{X}_{n_1-k_1+1} \\ \vdots & \vdots & \vdots & \vdots \\ \mathbf{X}_{k_1-1} & \mathbf{X}_{k_1} & \cdots & \mathbf{X}_{n_1-1} \end{bmatrix},$$

where each block is  $k_2 \times (n_2 - k_2 + 1)$  Hankel matrix:

$$\mathbf{X}_l = \begin{bmatrix} x_{l,0} & x_{l,1} & \cdots & x_{l,n_2-k_2} \\ x_{l,1} & x_{l,2} & \cdots & x_{l,n_2-k_2+1} \\ \vdots & \vdots & \vdots & \vdots \\ x_{l,k_2-1} & x_{l,k_2} & \cdots & x_{l,n_2-1} \end{bmatrix}.$$

# Multi-fold Hankel matrix (Hua '92)

---



$X_e$

# Low-rank structure of enhanced matrix

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- Enhanced matrix can be decomposed as

$$\mathbf{X}_e = \begin{bmatrix} \mathbf{Z}_L \\ \mathbf{Z}_L \mathbf{Y}_d \\ \vdots \\ \mathbf{Z}_L \mathbf{Y}_d^{k_1-1} \end{bmatrix} \text{diag}(\mathbf{d}) \left[ \mathbf{Z}_R, \mathbf{Y}_d \mathbf{Z}_R, \dots, \mathbf{Y}_d^{n_1-k_1} \mathbf{Z}_R \right],$$

- $\mathbf{Z}_L$  and  $\mathbf{Z}_R$  are Vandermonde matrices specified by  $z_1, \dots, z_r$
  - $\mathbf{Y}_d = \text{diag}[y_1, y_2, \dots, y_r]$
- Low-rank:  $\text{rank}(\mathbf{X}_e) \leq r$

# Recovery via Hankel matrix completion

---

Enhanced Matrix Completion (EMaC):

$$\begin{aligned} & \underset{\mathbf{z} \in \mathbb{C}^n}{\text{minimize}} && \|\mathbf{Z}_e\|_* \\ & \text{s.t.} && z_{i,j} = x_{i,j}, \quad (i,j) \in T \end{aligned}$$

- Can be easily extended to higher-dimensional frequency models



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